Dictionary Design for Graph Signal Processing

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Signal Processing on Graphs



Some Typical Graph Signal Processing Problems

Compression / Visualization





Denoising

Earth data source: Frederik Simons





Orthonormal Dictionaries



Orthonormal Dictionaries (cont.)





$$f = \sum_{\ell} \alpha_{\ell} u_{\ell} = \sum_{\ell} \langle f, u_{\ell} \rangle u_{\ell}$$

Overcomplete Dictionaries and Sparsity



- Given an overcomplete Φ , there are infinitely many choices of α that lead to the same signal f
- Useful to *sparsely* represent signals —> few non-zero coefficients in α

Motivating Example: Denoising

• Tikhonov regularization for denoising: $\operatorname{argmin}_{f} \left\{ ||f - y||_{2}^{2} + \gamma f^{T} \mathcal{L} f \right\}$



• Wavelet denoising: $\operatorname{argmin}_{a} \left\{ ||f - W^*a||_2^2 + \gamma ||a||_{1,\mu} \right\}$



Motiving Example: Compression

Piecewise-Smooth Signal with Discontinuities



Diffusion Wavelet Coefficients, Sorted by Magnitude





Dictionary Design for Signals on Graphs





Desirable Characteristics

- Ability to sparsely represent signals few non-zero coefficients in α
- Ability to capture the relevant characteristics of signals to extract information
- Computationally efficient to apply Φ and $\Phi^{\rm T}$
- Tight frames

Why Do We Need New Dictionaries?



To identify and exploit structure in the data, we need to account for the intrinsic geometric structure of the underlying graph data domain

The Essence of the Problem



- Weighted graphs are irregular structures that lack a shift-invariant notion of translation
- Many simple yet fundamental concepts that underlie classical signal processing techniques become significantly more challenging in the graph setting

Approach: Leverage Intuition from Euclidean Settings to Develop New Mathematical Tools for the Graph Setting



Generalized Operators

Combinatorial Graph Laplacian

- Connected, undirected, weighted graph $\mathcal{G} = \{\mathcal{V}, \mathcal{E}, W\}$
- Degree matrix D: zeros except diagonals, which are sums of weights of edges incident to corresponding node

Non-normalized graph Laplacian: $\mathcal{L} := \mathcal{D} - \mathcal{W}$

 Complete set of orthonormal eigenvectors and associated real, non-negative eigenvalues:

$$\mathcal{L}u_{\ell}=\lambda_{\ell}u_{\ell},$$

ordered w.l.o.g. s.t.

$$0 = \lambda_0 < \lambda_1 \leq \lambda_2 ... \leq \lambda_{N-1} := \lambda_{\max}$$

• Discrete difference operator: $(\mathcal{L}f)(i) = \sum_{i \in \mathcal{N}_i} W_{i,j}[f(i) - f(j)]$



Graph Fourier Transform

- Graph Laplacian eigenvectors are the analog of complex exponentials: Values of the eigenvectors associated with low eigenvalues change less rapidly across connected vertices
- Different choices of graph Fourier basis include combinatorial/normalized/random walk Laplacian eigenbasis or generalized eigenbasis of adjacency matrix



The GFT Incorporates the Graph Structure



The GFT Incorporates the Graph Structure



Graph Spectral Filtering

- Filtering: represent an input signal as a combination of other signals, and amplify or attenuate the contributions of some of the component signals
- In classical signal processing, the most common choice of basis the complex exponentials, which results in frequency filtering

$$f(t) \longrightarrow FT \longrightarrow \hat{f}(\xi) \longrightarrow \hat{g} \longrightarrow \hat{g}(\xi)\hat{f}(\xi) \longrightarrow IFT \longrightarrow \Phi f(t)$$

$$f(t)=20\cos(2\pi(1)t) + 2\cos(2\pi(11)t) \qquad \hat{f}(\xi) \qquad \hat{f}(\xi) \qquad \hat{f}(\xi) \qquad \hat{f}(\xi)\hat{g}(\xi) \qquad \Phi f(t)$$

$$\stackrel{20}{\xrightarrow{0}} \stackrel{10}{\xrightarrow{0}} \stackrel{10}{\xrightarrow{0}$$

Example: Image Denoising by Low-Pass Graph Filtering

$$f(n) \longrightarrow GFT \longrightarrow \hat{f}(\lambda_{\ell}) \longrightarrow \hat{g} \longrightarrow \hat{g}(\lambda_{\ell}) \hat{f}(\lambda_{\ell}) \longrightarrow IGFT \longrightarrow \Phi f(n)$$



Shuman et al., The emerging field of signal processing on graphs, SPM, 2013

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Approximating a Matrix Function Times a Vector

- Filtering: $g(\mathcal{L})f = Ug(\Lambda)U^*f$
- Too expensive to compute U and Λ for large graphs
- Common approach: estimate λ_{max} and approximate the filter g on the interval $[0, \lambda_{max}]$ by a polynomial, rational, or spline function
- Example: Truncated Chebyshev polynomial approximation

$$g(\mathcal{L})f = \frac{1}{2}c_0f + \sum_{k=1}^{\infty} c_k \bar{T}_k(\mathcal{L})f \approx \frac{1}{2}c_0f + \sum_{k=1}^{K} c_k \bar{T}_k(\mathcal{L})f =: \tilde{g}(\mathcal{L})f$$

• Use the three-term recurrence relation to compute $\bar{T}_k(\mathcal{L})f$ from

- $\overline{T}_{k-1}(\mathcal{L})f$ and $\overline{T}_{k-2}(\mathcal{L})f$, at the cost of one sparse matrix-vector multiplication by \mathcal{L}
- Pros: Fast for large, sparse graphs $[\mathcal{O}(K|\mathcal{E}|)]$; convergence guarantees when the filter g is analytic/smooth; distributable Druskin and Knizhnerman, "Two polynomial methods of calculating functions of symmetric matrices," 1989

Generalized Translation/Localization

- Define a generalized convolution by imposing that convolution in the vertex domain is multiplication in the graph spectral domain
- Define generalized translation via generalized convolution with a delta (i.e., filter a delta)

Functions on the Real Line

For $f \in L^2(\mathbb{R})$, in the weak sense $(T_s f)(t) := f(t - s)$ $= (f * \delta_s)(t)$ $= \int_{\mathbb{R}} \hat{f}(\xi) e^{-2\pi i \xi s} e^{2\pi i \xi t} d\xi$

Functions on the Vertices of a Graph

For $f \in \mathbb{R}^N$, we define $(T_i f)(n) := \sqrt{N}(f * \delta_i)(n)$ $= \sqrt{N} \sum_{\ell=0}^{N-1} \hat{f}(\lambda_\ell) u_\ell^*(i) u_\ell(n)$



Properties of Generalized Translation/ Localization

Warning 1: Do not have the group structure of classical translation:

 $T_i T_j \neq T_{i+j}$

Warning 2: Unlike the classical case, generalized translation operators are not unitary, so ||T_ig||₂ ≠ ||g||₂ in general

• However, the mean is preserved: $\sum_{n} (T_i g)(n) = \sum_{n} g(n)$

Theorem (Smoothness of \hat{g} leads to localization of $T_i g$ around vertex i)

Let $\hat{g} : [0, \lambda_{\max}] \to \mathbb{R}$ be a kernel and define $d_{in} := d_{\mathcal{G}}(i, n)$. Then

$$|(T_ig)(n)| \leq \sqrt{N}B_{\hat{g}}(d_{in}-1),$$

where $B_{\hat{g}}(K)$ is the minimax polynomial approximation error over all polynomials of degree K:

$$B_{\hat{g}}(K) := \inf_{\widehat{p_{K}}} \left\{ \sup_{\lambda \in [0, \lambda_{\max}]} |\hat{g}(\lambda) - \widehat{p_{K}}(\lambda)|
ight\}.$$

Downsampling and Graph Reduction

Downsampling



- Downsampling + graph reduction = a multiresolution of graphs
- Methods used here:
 - Graph downsampling by polarity of Laplacian eigenvector associated with largest eigenvalue
 - Kron reduction with spectral sparsification
- Alternative: coarse graining



Sampling and Interpolation

- How to sample a graph signal and interpolate from the samples?
- How to choose the samples depends on your prior knowledge of the data
- Subset V_s of vertices is a <u>uniqueness set</u> for a subspace P iff:
 - If two signals in the subspace P have the same values on the vertices in the uniqueness set, then they are the same signal



Can we recover all 500 values of this signal from 30 measurements? If so, where should we take those measurements?

Sampling and Interpolation: Signals Concentrated on Spectral Bands

Example: subspace of globally smooth signals with band limit λ_{29}



1. Recover graph Fourier coefficients:

$$U_{\mathcal{S},\mathcal{R}}x = f_{\mathcal{S}}$$

2. Interpolation / reconstruction: $\tilde{f} = U_{:,\mathcal{R}} x$





Approaches to Graph Signal Dictionary Design



Analytic Versus Trained Dictionaries

- Rubinstein et al., Dictionaries for sparse representation modeling, Proc. IEEE, 2010
- Analytic dictionaries: adapted to graph structure, but not Analytic dictionaries: adapted to graph structure, but not Analytic dictionaries: adapted to graph structure, but not
- Dictionary learning: adapt dictionary to training data
 - Aharon et al., The K-SVD, TSP, 2003
 - Engan et al., Method of optimal directions for frame design, ICASSP, 1999
 - These general methods do not explicitly account for graph structure
 - <u>Parametric training</u>: force some structure upon the dictionary (e.g., to incorporate graph topology, ensure an efficient computational implementation), but use training signals to learn parameters

Survey of Approaches to Graph Signal Dictionary Design

- Graph Fourier transform
- Vertex domain designs
- Diffusion-based designs
- Windowed graph Fourier transform
- Spectral domain designs
- Generalized filter banks

Motivating Example: Any Structure?



Classical Windowed Fourier Transform

- Localized Fourier analysis joint descriptions of signals' temporal and spectral behavior
 - Localized oscillations appear frequently in audio processing, vibration analysis, radar detection, etc.
 - e.g., identify musical notes and melody at different times
- Windowed (short-time) Fourier transform of $f \in L^2(\mathbb{R})$:

$$Sf(s,\xi) := \langle f, g_{s,\xi} \rangle = \int_{-\infty}^{\infty} f(t) \overline{g(t-s)} e^{-2\pi i \xi t} dt$$

នំ <u>នៅលេក្រម</u>្លាល់ ក្រក់កំ <u>ក្រ</u>ស់ នំ នៅ ស្រី ស្រី ស្រី ស្រី សំរាំ នំ លេក សំរាំ សំរាំ សំរាំ សំរាំ សំរាំ សំរាំ សំរាំ នំ លេក សំរាំ សំរាំង សំរាំ សំរាំ សំរាំ សំរាំង សំរាំ សំរាំ សំរាំ សំរាំ សំរាំ សំរាំ សំរាំ សំរាំ សំរាំ សំរាំង សំរាំង សំរាំ សំរាំ សំរាំ សំរាំ សំរាំ សំរាំង សំរាំង សំរាំ សំរាំង អាំង សំរាំង អាំង សំរាំង អាំង សំរាំង សំរាំង សំរាំង សំរាំង សំរាំង សំរាំង សំរាំង សំរាំង អាំង សំរា









Windowed Graph Fourier Transform

1 Translate a window g to each vertex of the graph



2 Multiply each component of the graph signal f of interest by the corresponding component of the translated window T_ig

Take the graph Fourier transform of $f \cdot * T_i g$ (recall analysis)

Shuman et al., Vertex-frequency analysis on graphs, ACHA, 2016

Windowed Graph Fourier Transform (cont.)

• Windowed graph Fourier atoms: $g_{i,k} := M_k T_i g$



Spectrogram Examples





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Spectrogram Examples

• Spectrogram = frequency-lapse video







Survey of Approaches to Graph Signal Dictionary Design

- Windowed graph Fourier transform
- Spectral domain designs
- Generalized filter banks

Dictionary of Localized Kernels

- M kernels/patterns $\{g_m\}$
- Localize each pattern to each vertex
 - Atoms of the form $T_i g_m = g_m(\mathcal{L}) \delta_i$

-
$$\Phi = [g_1(\mathcal{L}), g_2(\mathcal{L}), \dots, g_M(\mathcal{L})]$$

- Dictionary is overcomplete with MN atoms
- Approximate kernels with polynomials
 - Ensures joint localization in both domains
 - Fast computations with dictionary and its adjoint



Example 2 (Tikhonov regularization): We observe a noisy graph signal $\mathbf{y} = \mathbf{f}$ Gaussian noise, and wish to recover \mathbf{f}_0 . To enforce a priori information that the the underlying graph, we include a regularization term of the form $\mathbf{f}^{\mathrm{T}} \mathcal{L} \mathbf{f}$, and,

 $\operatorname{argmin}_{\mathbf{c}} \left\{ \|\mathbf{f} - \mathbf{y}\|_{2}^{2} + \gamma \mathbf{f}^{\mathrm{T}} \mathbf{f} \mathbf{f} \right\}.$

Example: Spectral Graph^{the underlying graph we include we include the underlying graph we include the underlying}

Hammond et al., Wavelets on graphs vithe proposition 1]) the optimal reconstruction is given by

• Generalized dilation:





Improvement 1: Energy Conservation



Leonardi and Van De Ville, Tight wavelet frames on multislice graphs, TSP, 2013
 Shuman et al., Spectrum-adapted tight graph wavelet and vertex-frequency frames, TSP, 2015

Improvement 2: Discrimination Power

- Ideally, atoms should not be too correlated with each other
- An extreme example:



• Cumulative coherence for a given sparsity level k

$$\mu_1(k) := \max_{|\Theta|=k} \max_{\psi \in \mathcal{D}_{\{1,2,\dots,N \in M\} \setminus \Theta}} \sum_{\theta \in \Theta} \frac{|\langle \psi, \mathcal{D}_{\theta} \rangle|}{||\psi||_2 ||\mathcal{D}_{\theta}||_2}$$

 $|/_{a}, \boldsymbol{\pi} \rangle|$

Aside: Fast Estimation of the Spectral Distribution with the Kernel Polynomial Method



Lin et al., "Approximating spectral densities of large matrices," SIAM Review, 2016

Spectral Distribution Examples



Improvement 2: Discrimination Power

• With access to a rough estimate of the spectral density, we can adapt the filters to the spectrum via warping



Shuman et al., Spectrum-adapted tight graph wavelet and vertex-frequency frames, TSP, 2015

Variant: Parametric Learning

• Restrict kernels to be polynomials of a given degree, and learn the polynomial coefficients from a training data set



Zhang et al., Learning of structured graph dictionaries, ICASSP, 2012Thanou et al., Learning parametric dictionaries for signals on graphs, TSP, 2014

Survey of Approaches to Graph Signal Dictionary Design

- Windowed graph Fourier transform
- Spectral domain designs
- Generalized filter banks

1D Wavelets Via Filter Banks

Classical 2-Channel Critically Sampled Filter Bank



- To extend to the graph setting, we need appropriate notions of downsampling, upsampling, filtering, graph reduction
- Some issues that arise:
 - Difficulty generalizing conditions on filters ensuring properties such as perfect reconstruction, orthogonality
 - Preserving a meaningful correspondence between filtering at different resolution levels

Iterating Low Pass Branch Yields Wavelets



M-Channel Critically Sampled Graph Filter Bank

Architecture



Jin and Shuman., "An M-channel critically sampled filter bank for graph signals," ICASSP, 2017

M-Channel Critically Sampled Graph Filter Bank

• Partition into uniqueness sets for ideal filter bank subspaces:



• To avoid a full eigendecomposition, we would like to use random, non-uniform sampling and fast, approximate reconstruction methods

Greene and Magnanti, "Some abstract pivot algorithms," SIAM J. Appl. Math., 1975

Non-Uniform Random Sampling



Non-Uniform Random Sampling



Puy et al., "Random sampling of band limited signals on graphs," ACHA, 2016



Signal-Adapted M-CSFB





Also reallocate samples across bands based on signal's energy

Li, Jin, and Shuman, "Scalable M-channel critically sampled filter banks," 2018

Joint Localization of Atoms

- Dictionary atoms are of the form $\tilde{h}_m(\mathcal{L})\delta_i$
- Localized within K hops of center vertex
- As K increases, become more concentrated in spectral domain





nterpolate nterpolate **Efficient Interpolation** downsampling operator $\min_{\substack{\mathcal{O} \mid (U, \tau_{m})}} ||\Omega_{m, \mathcal{V}_{m}}^{-1/2} (M_{m} z - y_{\mathcal{V}_{m}})||_{2}^{2}$ approximate by convex $z \in \operatorname{col}(U_{:,\mathcal{R}_m})$ optimization. problem signal model space $\min_{z \in \mathbb{R}^N} \left\{ \gamma || \Omega_{m, \mathcal{V}_m}^{-1/2} (M_m z - y_{\mathcal{V}_m}) ||_2^2 + z^\top \varphi_m(\mathcal{L}) z \right\}$ optimality condition $\left(\gamma M_m^{\top} \Omega_{m,\mathcal{V}_m}^{-1} M_m + \varphi_m(\mathcal{L})\right) z = \gamma M_m^{\top} \Omega_{m,\mathcal{V}_m}^{-1} y_{\mathcal{V}_m}$

solve with preconditioned conjugate gradient preconditioner:

Li, Jin, and Shuman, "Scalable M-channel critically sampled filter banks for graph signals," 2018

Puy et al., "Random sampling of band limited signals on graphs," ACHA, 2016

Compression Example

Original Signal



- N = 469,404
- Computation times
 - Analysis: 45-90 sec
 - Synthesis: 90-1000 sec

Reconstruction from 10% of Coefficients



Reconstruction Error

5

0

-5

-8



Recent, Ongoing, and Future Work



Explore



- <u>https://lts2.epfl.ch/gsp/</u>
- <u>https://www.macalester.edu/~dshuman1/publications.html</u>